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A statistical approach to the optimization of the radar ambiguity function and the chaos-based waveform design

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Abstract

In this paper we adopt a statistical approach to optimize the ambiguity function of a radar system. By considering the codes defining the transmitted waveform as realizations of a random variable we firstly show that a suitable distribution of the random variable allows to obtain good codes. Secondly we show that using the chaotic skew tent map it is possible to generate deterministic codes having the desired statistical properties. This allows to obtain an optimized global ambiguity function of the radar system. The advantage of using chaos-based sequences is that they can be easily generated in any length and number. We further improve their performance by introducing down sampling. It appears that the proposed sequences have performance quite similar to those of the sequences of the literature computationally optimized.

Keywords: Radar ambiguity function, Waveform design, Statistical approach, Chaos-based sequences, Skew tent map, Invariant probability density

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1. Introduction

The primary purpose of a radar system [1] is to extract information about potentially moving targets in a given propagation environment by transmitting well-chosen waveforms and analysing the signals returned to the radar after their reflection on the targets. In particular the parameters of interest are the range and the speed of the target which can be measured by the round trip time and the Doppler frequency shift of the signal received by the radar, respectively. The system performance is based both on the receive filter and transmit waveform. We here focus on waveform design. The ambiguity function [2,3] describes the response of a matched filter to a signal for different time delays and Doppler frequencies. The search for transmitted waveforms leading to the "best possible" radar ambiguity function, especially with regard to the low side lobes, has already been addressed in the literature [1]. In [5,6] the authors presented the problem of waveform design as the nonlinear optimization of the radar ambiguity function and proposed algorithms which are relatively expensive in computation and / or complex to implement.

Waveform optimization has also recently sparked renewed interest in the context of MIMO radar [7] which consists of multiple antenna elements transmitting different waveforms toward different angles. In this context some works of the literature are inspired by the field of multi-user communications [8,9], others aim to synthesize sequences by optimizing certain criteria concerning their auto- and cross-correlation functions [10-14]. It however appears that the waveforms generated by the existing methods have drawbacks. Either they are limited in length, or they require important calculations, especially when we need a large number of them or we want to add one.

Because the optimization of the global ambiguity function, that is to say over all time delays and Doppler frequencies, is quite difficult, other works [15-17] suggested locally optimizing the ambiguity function over a given range of time delays and Doppler frequencies. Our approach proposed here considers the global ambiguity function.

In this paper, we propose to consider chaotic sequences as an alternative to other sequences in the literature for the design of radar waveforms. Earlier articles in the context of multi-user CDMA communication have already shown the interest of using codes based on chaos over more traditional codes [18,19].

35 Some other works have also suggested the use of chaotic sequences as candidates for the design of radar waveforms (see [20-22] and the references inside). In this paper we propose a statistical approach to exploit the features of chaos in radar system. By considering each code as a realization of a random variable we show that we can build good codes if the distribution of the random variable is

40 suitably chosen. We then consider sequences generated by the skew tent map and we show that for some values of the bifurcation parameter the invariant probability density coincides with the desired one. Note that this approach is valid for all the values of the Doppler frequency and thus allows the optimization of the global ambiguity function.

45 After briefly introducing in section 2 the ambiguity function and extracting the function of interest to be optimized, we will study in section 3 the statistical properties of the latter. In section 4, we will show how sequences generated by the skew tent map meet the statistical properties established in the previous section and required for a good ambiguity function. We will also propose a down-

50 sampling of our sequences in order to further improve their behaviour which will be compared to computationally optimized sequences of the literature. Finally the conclusion will summary the contribution of the paper and will present future extensions.

2. Short reminder of the radar ambiguity function

55 The ambiguity function (AF) of a radar system consists of the 2D output, for a given time delay τ and a given Doppler frequency ν , of the filter matched to the transmitted signal $s(t)$ and can be written as [7]:

$$A(\tau, \nu) = \int s(t)s^*(t + \tau)e^{j2\pi\nu t} dt \quad (1)$$

where the waveform $s(t)$ is given by:

$$s(t) = \sum_{p=1}^{N_c} w_p u(t - (p-1)T_c) \quad (2)$$

N_c is the length of the sequences $\{w_p\}_{p=1, N_c}$ and $u(t)$ is a shaping function
of duration T_c . (1) then becomes :

$$A(\tau, \nu) = \sum_{p=1}^{N_c} \sum_{l=1}^{N_c} w_p w_l^* \tilde{\gamma}_{p,l}^u(\tau, \nu) \quad (3)$$

where

$$\tilde{\gamma}_{p,l}^u(\tau, \nu) = \int u(t - (p-1)T_c) u(t - (l-1)T_c + \tau) e^{j2\pi\nu t} dt \quad (4)$$

After some calculations the ambiguity function for $\tau = kT_c$ then becomes:

$$A(kT_c, \nu) = R_w(\nu, k) \alpha(\nu) \quad (5)$$

where

$$\alpha(\nu) = \int_0^{T_c} |u(t)|^2 e^{j2\pi\nu t} dt \quad (6)$$

and

$$R_w(k, \nu) = \sum_{p=1}^{N_c-k} \omega_p^* \omega_{p+k} e^{-j\pi\nu(p-1)T_c} \quad (7)$$

Note that in the case where $u(t)$ is the rectangular function of support $[0, T_c]$,

$$\alpha(\nu) = e^{j\pi\nu T_c} \frac{\sin \pi\nu T_c}{\pi\nu} \quad (8)$$

In practice, the Doppler frequency ν is usually much smaller than the bandwidth of the probing waveform so that we can safely suppose that $|\frac{\sin \pi\nu T_c}{\pi\nu}| \simeq T_c$. It then appears that the optimization of the ambiguity function reduces to the optimization of $|R_w(k, \nu)|$, it is to say that $|R_w(k, \nu)|$ for $k \neq 0$ must be the
smallest as possible compared to $|R_w(0, 0)|, \forall \nu$.

We here consider phased codes so that $\omega_p = e^{j\pi x_p}$, where x_p is a sequence in the interval $[-1, 1]$. The modulus of the expression (7) (we omit the subscript w in the following) becomes

$$|R(k, \nu)| = \left| \sum_{p=1}^{N_c-k} e^{j\pi z_p(k)} \right| \quad (9)$$

where

$$z_p(k) = x_{p+k} - x_p - \nu(p-1)T_c \quad (10)$$

75 If the sequence x_p is randomly generated, $z_p(k)$ is also a random variable for every fixed integer $k \neq 0$ and Doppler frequency ν . Note that for $k = 0$, $|R(0, \nu)| = \left| \frac{\sin(\pi\nu T_c N_c/2)}{\sin(\pi\nu T_c/2)} \right|$ so that $|R(0, 0)| = N_c$.

In the section below we will first analyse the statistics of the modulus $R =$
80 $|Z_n|$, for a fixed n , with

$$Z_n = \sum_{p=1}^n e^{j\pi y_p} \quad (11)$$

where n is large enough and y_p is a random variable. We will first establish the desired statistical properties for the random variable $R = |Z_n|$, in order to minimize its mean and variance which are related to the side lobes of the ambiguity function. We will then consider different distributions for y_p in (11).
85 The optimisation of $R = |Z_n|$ through the statistics of y_p will then be useful to the choice of the sequences x_p allowing the desired properties of z_p in (9) and (10).

We will also analyse the influence of the deterministic term in (10) corresponding to the Doppler frequency ν .

90 **3. Statistical properties of $R = |Z_n|$**

Let y_p be an *i.i.d* random variable. By noting α and β the mean values of the real and imaginary parts of Z_n , and by s_1 and s_2 their variances respectively we introduce the following notations as in [23]

$$\begin{cases} X = \frac{R}{\sqrt{s_1 + s_2}} \\ B = \frac{\alpha}{\sqrt{s_1 + s_2}} \\ K = \sqrt{\frac{s_2}{s_1}} \end{cases} \quad (12)$$

In the case of symmetric distribution with respect to 0 (i.e. $\beta = 0$) the distribution of the random variable X is [23]
95

$$f_X(x) = \frac{K^2 + 1}{K} x \exp\left[-\frac{K^2 + 1}{2} \left(B^2 + \frac{K^2 + 1}{2K^2} x^2\right)\right]$$

$$\sum_{m=0}^{\infty} (-1)^m \varepsilon_m I_m\left(\frac{K^4 - 1}{4K^2} x^2\right) I_{2m}[B(1 + K^2)x] \quad (13)$$

where

$$\varepsilon_m = \begin{cases} 1, & \text{if } m = 0; \\ 2, & \text{if } m \neq 0. \end{cases}$$

I_m is the modified Bessel function of the first kind.

In [23] it has been shown that:

- by varying K the distribution is only slightly modified, so we will consider $K = 1$ in the following.

100

- the mean and the variance are the smallest for $B = 0$, corresponding to $\alpha = 0$.

The distribution (13) is then almost optimal in the sense that for a given value x_0 of X the probability $P(X < x_0)$ is the greatest for $B = 0$. This is illustrated in Figure 1 for $K = 1, 2$ and $B = 0, 0.5, 1, 2$.

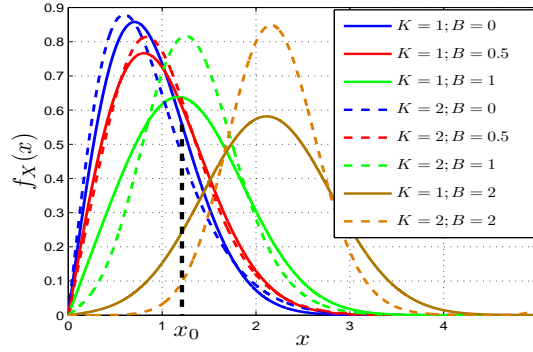


Figure 1: Distribution $f_X(x)$ for different values of K and B

105

For $K = 1$ the distribution (13) is then reduced to

$$f_X(x) = 2xe^{-(B^2+x^2)} I_0(2Bx) \quad (14)$$

Using (12) we retrieve the distribution of the random variable $R = |Z_n|$

$$f_R(r) = 2 \frac{r}{s_1 + s_2} e^{-(B^2 + \frac{r^2}{s_1 + s_2})} I_0\left(2B \frac{r}{\sqrt{s_1 + s_2}}\right) \quad (15)$$

- For $\alpha \neq 0$ and thus $B \neq 0$ the distribution (15) is the Rice distribution
- For $\alpha = 0$ and thus $B = 0$ we obtain the Rayleigh distribution

$$f_R(r) = 2 \frac{r}{s_1 + s_2} e^{-\frac{r^2}{s_1 + s_2}} \quad (16)$$

110 The mean value and the variance of the distribution (16) are:

$$\begin{cases} E_R = \frac{\sqrt{(s_1 + s_2)\pi}}{2} \\ V_R = (1 - \frac{\pi}{4})(s_1 + s_2) \end{cases} \quad (17)$$

This distribution will then be considered as a reference, i.e. as said above, a sequence y_p in (11) yielding to such a Rayleigh distribution will be considered as optimal.

In the following we will consider several distributions for y_p . First note that
115 since y_p is supposed to follow an *i.i.d* distribution we have

$$\alpha = nE[\cos \pi y_p] \quad (18)$$

3.1. Case y_p uniformly distributed in $[-a, a]$; $a \leq 1$

In this case

$$\beta = 0; \quad \alpha = n \text{sinc}(a) \quad (19)$$

where $\text{sinc}(x) = \frac{\sin(\pi x)}{\pi x}$.

$$\begin{aligned} s_1 &= \frac{n}{2}(1 + \text{sinc}(2a) - 2\text{sinc}^2(a)) \\ s_2 &= \frac{n}{2}(1 - \text{sinc}(2a)) \end{aligned}$$

120 which gives

$$\begin{cases} s_1 + s_2 = n(1 - \text{sinc}^2(a)) \\ B = \sqrt{n} \frac{\text{sinc}(a)}{\sqrt{1 - \text{sinc}^2(a)}} \end{cases} \quad (20)$$

The Rayleigh distribution (16) is obtained for a an integer. Indeed in this case $\text{sinc}(a) = 0$ which gives $B = 0$, $K = 1$ and $s_1 + s_2 = n$. Figure 2 exhibits an example of $f_R(r)$ for $n = 1023$ and for different values of a . In the case of the Rayleigh distribution (16) and from (17) the mean value and variance of

125 $R = |Z_n|$ are

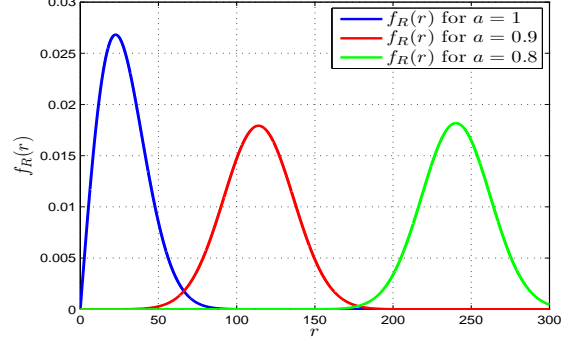


Figure 2: The distribution $f_R(r)$ for $n = 1023$ and $a = 1; 0.9; 0.8$

$$\begin{cases} E_R = \frac{\sqrt{n\pi}}{2} \\ V_R = (1 - \frac{\pi}{4})n \end{cases} \quad (21)$$

Note that, in practice, these mean and variance values should be compared to the maximum peak of the ambiguity function, i.e. $|R(0,0)| = n$, and its square value, respectively. Thus we considered the two more significant features: the normalized mean $E_n = \frac{E_R}{n}$ and variance $V_n = \frac{V_R}{n^2}$, we obtain

$$\begin{cases} E_n = \frac{1}{2} \sqrt{\frac{\pi}{n}} \\ V_n = (1 - \frac{\pi}{4}) \frac{1}{n} \end{cases} \quad (22)$$

130 It then appears that the mean side lobe to maximal peak ratio of the ambiguity
function vanishes as n increases as illustrated in Figure 3.

3.2. Case y_p uniformly distributed in $[-a, a]$; $a > 1$

Due to the fact that $e^{j\pi y_p}$ is 2-periodic, a random variable \tilde{y}_p in $[-1, 1]$
satisfying $e^{j\pi y_p} = e^{j\pi \tilde{y}_p}$ corresponds to the random variable y_p , see the example
135 plotted in red line on Figure 4. Consequently, the statistical properties of $|Z_n|$
are the same for y_p and \tilde{y}_p . It is easy to show that \tilde{y}_p is given by:

$$\tilde{y}_p = \begin{cases} y_p - \lfloor y_p \rfloor - 1 & \text{if } \lfloor y_p \rfloor \text{ is odd} \\ y_p - \lfloor y_p \rfloor & \text{if } \lfloor y_p \rfloor \text{ is even} \end{cases} \quad (23)$$

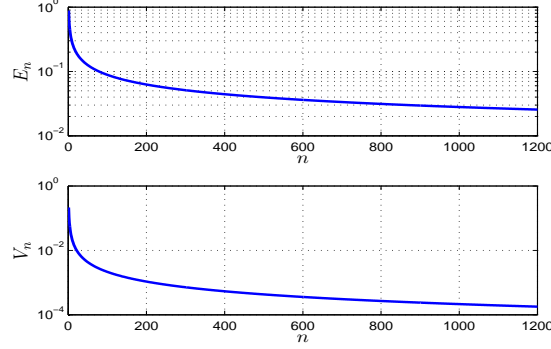


Figure 3: Mean and variance of the Rayleigh distribution versus sequence length n

$\lfloor x \rfloor$ is the floor of x . Let y_p be uniformly distributed in $[-a, a]$ where $a > 1$. We show in the appendix that $f_{\tilde{y}_p}(y)$ has two expressions according to $\lfloor a \rfloor$.

- if $\lfloor a \rfloor$ is even, $\lfloor a \rfloor = 2l$

$$f_{\tilde{y}_p}(y) = \begin{cases} \frac{l}{a} & \text{if } a - 2l \leq |y| \leq 1 \\ \frac{2l+1}{2a} & \text{if } 0 \leq |y| \leq a - 2l \end{cases} \quad (24)$$

- 140 • if $\lfloor a \rfloor$ is odd, $\lfloor a \rfloor = 2l + 1$

$$f_{\tilde{y}_p}(y) = \begin{cases} \frac{2l+1}{2a} & \text{if } 0 \leq |y| \leq -a + 2l + 2 \\ \frac{l+1}{a} & \text{if } -a + 2l + 2 \leq |y| \leq 1 \end{cases} \quad (25)$$

In Figure 4, we plotted the probability density of $f_y(y)$ and $f_{\tilde{y}}(y)$ in the two cases $\lfloor a \rfloor$ odd ($\lfloor a \rfloor = 5$) and $\lfloor a \rfloor$ even ($\lfloor a \rfloor = 6$). By an easy computation of $\alpha = \int_{-1}^1 \cos(\pi y) f_{\tilde{y}}(y) dy$ we found that

$$\alpha = \begin{cases} -\frac{\sin(\pi(2l+2-a))}{2\pi a} & \text{if } \lfloor a \rfloor = 2l + 1 \\ \frac{\sin(\pi(a-2l))}{2\pi a} & \text{if } \lfloor a \rfloor = 2l \end{cases} \quad (26)$$

In both cases $2l + 2 - a$ and $a - 2l \in [0, 1[$, $\alpha = 0$ if and only if a is an integer
145 corresponding to the uniform distribution in $[-1, 1]$. In other words we obtain R with the Rayleigh distribution (16) if and only if $\alpha = B = 0$.

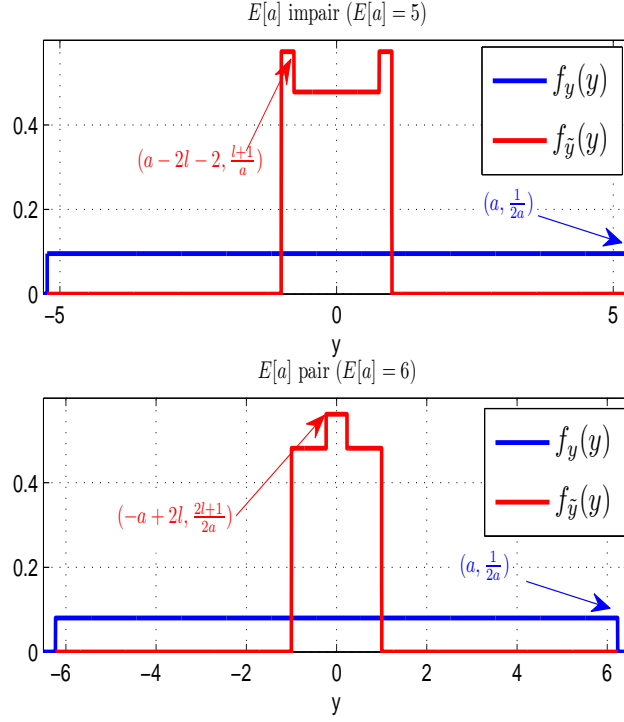


Figure 4: $f_y(y)$ and $f_{\tilde{y}}(y)$ in the case of uniform distribution.

3.3. Case of y_p with a triangular distribution in $[-a, a]$

First note that this distribution is obtained when $y_p = z_p(k)$ in (10) with $\nu = 0$ and uniformly distributed x_p and x_{p+k} .

150 We show in the appendix that $f_{\tilde{y}}(y)$ can have two expressions according to $\lfloor a \rfloor$:

- if $\lfloor a \rfloor$ is even, $\lfloor a \rfloor = 2l$

$$f_{\tilde{y}}(y) = \begin{cases} \frac{(2l+1)a - 2l(l+1) - |y|}{a^2} & \text{if } 0 \leq |y| \leq a - 2l \\ \frac{2la - 2l^2}{a^2} & \text{if } a - 2l \leq |y| \leq 1 \end{cases} \quad (27)$$

- if $\lfloor a \rfloor$ is odd, $\lfloor a \rfloor = 2l + 1$

$$f_{\tilde{y}}(y) = \begin{cases} \frac{(2l+1)a - 2l(l+1) - |y|}{a^2} & \text{if } 0 \leq |y| \leq 2l + 2 - a \\ \frac{2(l+1)a - 2(l+1)^2}{a^2} & \text{if } 2l + 2 - a \leq |y| \leq 1 \end{cases} \quad (28)$$

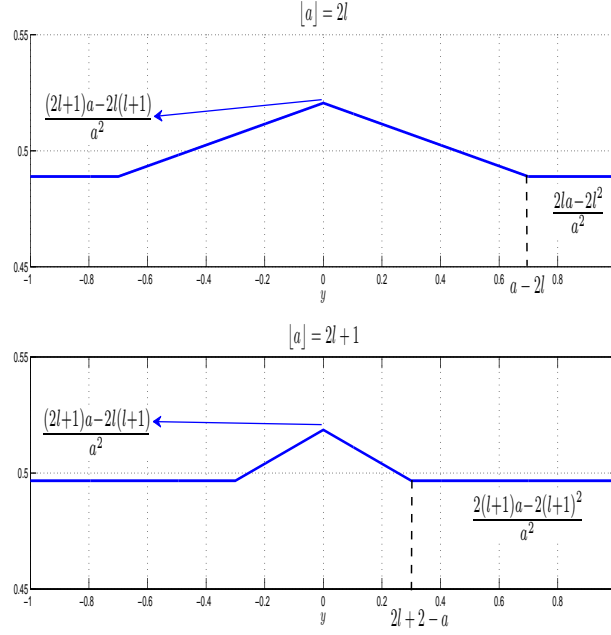


Figure 5: $f_{\tilde{y}}(y)$ when y_p has a triangular distribution in $[-a, a]$ in the cases a even ($a = 2l$) and a odd ($a = 2l + 1$)

In Figure 5 we plotted this distribution in both cases. Note that:

- for $[a]$ even, when a is an integer, i.e. $a = 2l$ we obtain the uniform distribution in $[-1, 1]$.
- for $[a]$ odd, when a is an integer, i.e. $a = 2l + 1$ we obtain the triangular distribution in $[-1, 1]$.

These two results explain the expressions of α obtained by the computation of $\alpha = \int_{-1}^1 \cos(\pi y) f_{\tilde{y}}(y) dy$ and given below

$$\alpha = \begin{cases} -2 \frac{1 - \cos(\frac{\pi(a-2l)}{\pi^2 a^2})}{\pi^2 a^2} & \text{if } [a] = 2l \\ 2 \frac{1 - \cos(\frac{\pi(2l+2-a)}{\pi^2 a^2})}{\pi^2 a^2} & \text{if } [a] = 2l + 1 \end{cases} \quad (29)$$

When $a = 2l$ we obtain $\alpha = 0$ giving the Rayleigh distribution (16), when $a = 2l + 1$ the expression (29) is maximal and equal to $\frac{4}{\pi^2 a^2}$ corresponding to a

Rice distribution (15).

3.4. Case of y_p centered and gaussian distributed

This distribution is obtained when x_p and x_{p+k} in (10) are independent gaussian and centred variables and $\nu = 0$. When $f_y(y)$ is gaussian, $f_{\tilde{y}}(y)$ has

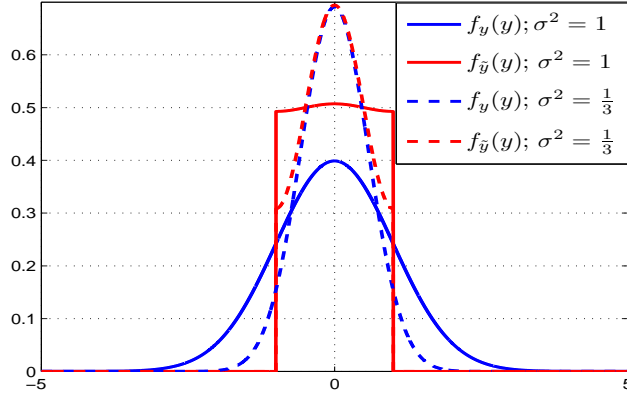


Figure 6: $f_y(y)$ et $f_{\tilde{y}}(y)$ in the case of centered and gaussian distribution

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one of the two shapes represented in Figure 6, according to the variance σ^2 :

- $\sigma < 1$ we obtain a density that differs very much from the uniform one, (as for $\sigma^2 = \frac{1}{3}$ in the figure).
- $\sigma \geq 1$, the distribution $f_{\tilde{y}}(y)$ is almost uniform in $[-1, 1]$.

170 3.5. Histograms of $|Z_n|$

To confirm the results of the discussion above we plotted in Figure 7 the histograms of $|Z_n|$ for the previously considered distributions with the Rayleigh distribution in red line . These histograms are computed for the case $n = 1023$ using 10000 realizations of y_p . We obtain a Rayleigh distribution of $R = |Z_n|$

175 when y_p follows the following distributions:

- uniform distribution in $[-q, q]$, q an integer;

- triangular distribution in $[-q, q]$, $q \geq 2$ an even integer;
- centered and gaussian distribution with variance greater or equal to 1;

For these results we can see that the average and the variance are very close
 180 to the theoretical values computed using (22) that are in the case $n = 1023$:
 $E_n \cong 28.3454, V_n \cong 219.5377$. In the other cases, we obtain a distribution
 which is very different from the Rayleigh distribution.

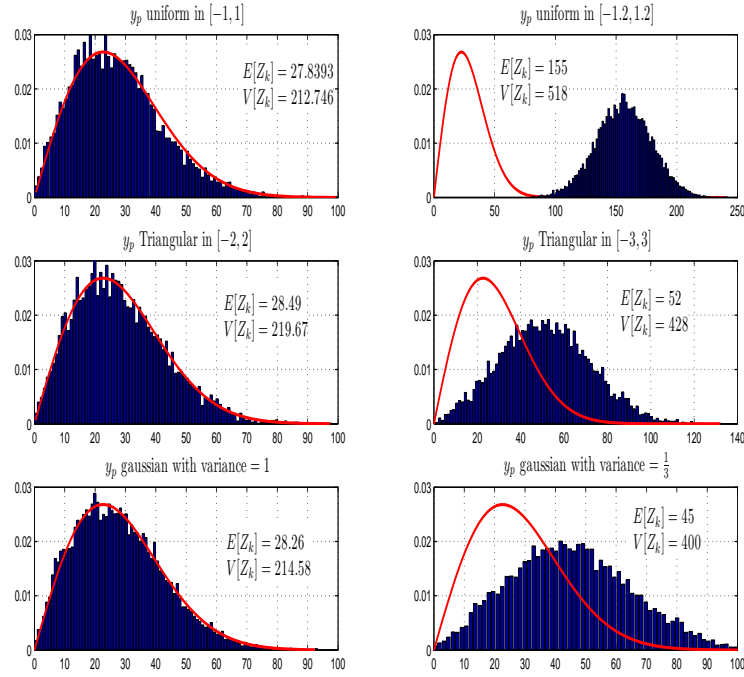


Figure 7: Histograms of Z_k for different distribution of y_p

3.6. Statistics of $R = |Z_n|$ in the case $y_p = u_p + \nu(p-1)T_c$

In this section we tackle the influence of the presence of the determinisic
 185 term $\nu(p-1)T_c$ on the statistics of R . We consider $y_p = u_p + \nu(p-1)T_c$ where
 u_p are i.i.d. random variables. It is easy to see that as the u_p are i.i.d. $E[e^{j\pi u_p}]$

is a constant (not depending on p) denoted by A and

$$E[Z_n] = A \sum_{p=1}^n e^{j\pi(p-1)\nu T_c} \quad (30)$$

In the case $\nu T_c = 2q$ where q is an integer, $E[Z_n] = An$. If $A = 0$ the real part of Z_n is also zero and consequently $\alpha = 0$ and $B = 0$ which are the conditions so that $R = |Z_n|$ has a Rayleigh type distribution (see [23]).

In case $\nu T_c \neq 2q$:

$$E[Z_n] = A \frac{e^{j\pi n \nu T_c / 2} \sin[\pi n \nu T_c / 2]}{e^{j\pi \nu T_c / 2} \sin[\pi \nu T_c / 2]} \quad (31)$$

Since $|\sin[\pi n \nu T_c / 2]| \leq n |\sin[\pi \nu T_c / 2]|$, we obtain

$$|\text{Real}(E[Z_n])| \leq |E[Z_n]| \leq |A|n \quad (32)$$

Also in this case if $A = 0$, the real part of $E[Z_n]$ is zero, thus $\alpha = 0$ and $B = 0$ which are the conditions so that $R = |Z_n|$ has a Rayleigh type distribution.

To illustrate this idea we plotted on Figure 8 the histograms of $|Z_n|$ for y_p uniform in $[-1.2, 1.2]$. We can see that the distribution of $|Z_n|$ differs from the desired Rayleigh one only when $\nu = 0$. This could be explained by saying that the presence of the deterministic term allows uniform scattering of the points on the unit circle when the distribution of u_p is not uniformly distribution in $[-1, 1]$.

Thus the presence of the deterministic term $\nu(p-1)T_c; \nu \neq 0$ makes every sequence a good one since it allows scattering the points on the unit circle. We can conclude that the important thing is to optimize $|Z_n|$ for $\nu = 0$.

4. Ambiguity function for chaotic sequences

In this section we consider chaotic sequences $\{x_p\}$ generated by $x_{p+1} = T_\mu(x_p)$ and an initial condition $x_0 \in [-1, 1]$ where $\mu \in [-1, 1]$; $T_\mu(x)$ is the piece-wise linear skew tent map defined in $[-1, 1]$ by

$$T_\mu(x) = \begin{cases} \frac{2}{\mu-1}x - \frac{1+\mu}{\mu-1} & \text{if } \mu < x \leq 1 \\ \frac{2}{\mu+1}x - \frac{\mu-1}{\mu+1} & \text{otherwise} \end{cases} \quad (33)$$

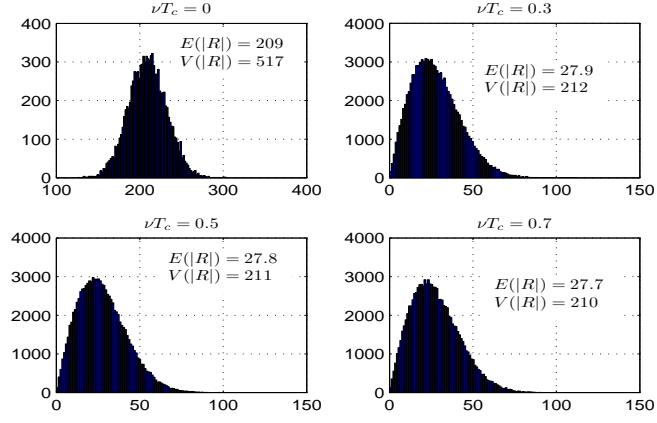


Figure 8: Histograms of $|Z_k|$ in the case of y_p uniform in $[-1.2, 1.2]$

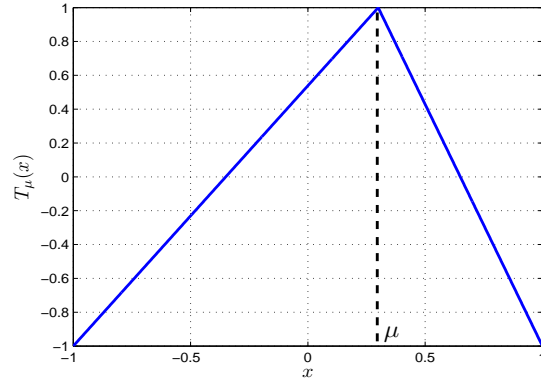


Figure 9: Curve of $T_\mu(x)$

The curve of $T_\mu(x)$ is given on Figure 9. The invariant probability density of the variable x_p is the uniform distribution in the interval $[-1, 1]$ [24] i.e. for N large enough $x_p, p = 0, 1, \dots, N$ could be considered as a realization of a random variable uniformly distributed in $[-1, 1]$. The idea is to use such sequences in the radar system described above with $y_p = g_k(x_p)$ for a given integer k where

$$g_k(x) = T_\mu^k(x) - x \quad (34)$$

To analyze the statistics of y_p for a given k , we plotted the curve of $y = g_k(x)$ in Figure 4 for $k = 4$ (red line). The argument and the value of the maxi-

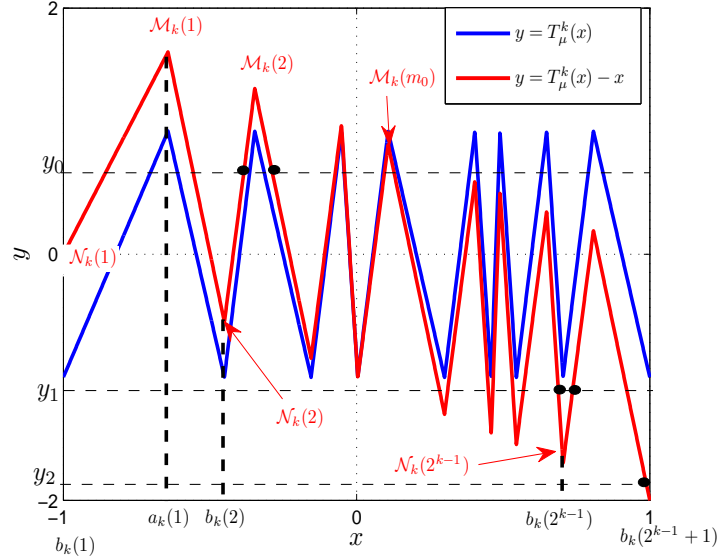


Figure 10: Curves of $y = g_k(x)$; $k = 4$

mum and the minimum are respectively $(a_k(m), \mathcal{M}_k(m)), 1 \leq m \leq 2^{k-1}$ and $(b_k(m), \mathcal{N}_k(m)), 1 \leq m \leq 2^{k-1} + 1$ where

$$\mathcal{M}_k(m) = 1 - a_k(m), 1 \leq m \leq 2^{k-1} \quad (35)$$

$$\mathcal{N}_k(m) = -1 - b_k(m), 1 \leq m \leq 2^{k-1} + 1. \quad (36)$$

$a_k(m)$ and $b_k(m)$ are defined by the following recursive relations[25]:

$$a_1(1) = \mu, \quad b_1(1) = -1, \quad b_1(2) = 1$$

$$a_{k+1}(2m-1) = \frac{\mu+1}{2}[a_k(m) - b_k(m)] + b_k(m)$$

$$a_{k+1}(2m) = \frac{\mu+1}{2}[a_k(m) - b_k(m+1)] + b_k(m+1)$$

$$b_{k+1}(2m-1) = b_k(m)$$

$$b_{k+1}(2m) = a_k(m)$$

$$b_{k+1}(2m+1) = b_k(m+1)$$

220

The curve of $T_\mu^k(x) - x$ consists of alternately 2^{k-1} increasing and decreasing
225 straight lines. Let $y = T_\mu^k(x) - x$ it is easy to see that

- in the m^{th} increasing line we have

$$x = A_k(m)y + B_k(m) \tag{37}$$

$$A_k(m) = \frac{a_k(m) - b_k(m)}{2 - a_k(m) + b_k(m)}$$

$$B_k(m) = \frac{a_k(m) + b_k(m)}{2 - a_k(m) + b_k(m)}$$

- in the m^{th} decreasing line we have

$$x = C_k(m)y + D_k(m) \tag{38}$$

230

$$C_k(m) = \frac{a_k(m) - b_k(m+1)}{2 - a_k(m) + b_k(m+1)}$$

$$D_k(m) = \frac{a_k(m) + b_k(m+1)}{2 - a_k(m) + b_k(m+1)}$$

Let X be a random variable following the uniform distribution in $[-1, 1]$ and
 $Y = T_\mu^k(X) - X$. The probability $P(Y < y) = P(X \in g_k^{-1}(]-2, y])$. Since X
is supposed to be uniformly distributed in $[-1, 1]$ we can confirm the following
235 points about the distribution of Y .

- if $y \leq -2$ the probability $P(Y < y) = 0$.

- if $y \geq \mathcal{M}_k(1)$ the probability $P(Y < y) = 1$.
- if $0 \leq y_0 \leq \mathcal{M}_k(1)$ as shown in the figure there exists an unique integer m_0 , $1 \leq m_0 \leq 2^{k-1} - 1$ such that $y_0 \in [\mathcal{M}_k(m_0 + 1), \mathcal{M}_k(m_0)]$. Using (37) and (38) we can show easily that

$$P(Y < y_0) = \frac{1}{2} \left(2 - \sum_{m=1}^{m_0} [C_k(m) - A_k(m)] y_0 + [D_k(m) - B_k(m)] \right) \quad (39)$$

- if $\mathcal{N}_k(2^{k-1}) \leq y_1 \leq 0$ there exists an integer m_1 , $1 \leq m_1 \leq 2^{k-1} - 1$ such that $y_1 \in [\mathcal{N}_k(m_1 + 1), \mathcal{N}_k(m_1)]$. In this case

$$P(Y < y_1) = \frac{1}{2} \sum_{m=1}^{m_1} [(A_k(m) - C_k(m)) y_1 + B_k(m) - D_k(m)] + \frac{1}{2} [2 - (C_k(2^{k-1}) y_1 + D_k(2^{k-1}))] \quad (40)$$

- $-2 \leq y_2 \leq \mathcal{N}_k(2^{k-1})$

$$P(Y < y_2) = \frac{1}{2} (2 - (C_k(2^{k-1}) y_2 + D_k(2^{k-1}))) \quad (41)$$

By deriving the function $P(Y < y)$ in all the cases above we obtain the probability density $f_{\mu,k}(y)$ of $Y = T_{\mu}^k(X) - X$

$$f_{\mu,k}(y) = \begin{cases} \frac{\sum_{m=1}^{m_y} \gamma(k,m)}{2}, & \text{if } 0 \leq y \leq \mathcal{M}_k(1) \\ \frac{\sum_{m=1}^{m_y} \gamma(k,m) + C_k(2^{k-1})}{2}, & \text{if } \mathcal{N}_k(2^{k-1}) \leq y \leq 0 \\ \frac{-C(k, 2^{k-1})}{2} & \text{if } -2 < y \leq \mathcal{N}_k(2^{k-1}) \\ 0 & \text{otherwise} \end{cases} \quad (42)$$

where m_y is the integer satisfying $y \in [\mathcal{M}_k(m_y + 1), \mathcal{M}_k(m_y)]$ if $0 \leq y \leq 1 - a_k(1)$ or satisfying $y \in [\mathcal{N}_k(m_y + 1), \mathcal{N}_k(m_y)]$ if $-1 - b_k(2^{k-1}) \leq y < 0$, and

$$\gamma(k, m) = A(k, m) - C(k, m)$$

Figure 11 exhibits the probability density of $y = T_{\mu}^k(x) - x$ in blue line for $\mu = 0.7$ and for different values of k . In red line we plotted the probability density of \tilde{y} corresponding to y as defined in the previous section. From the

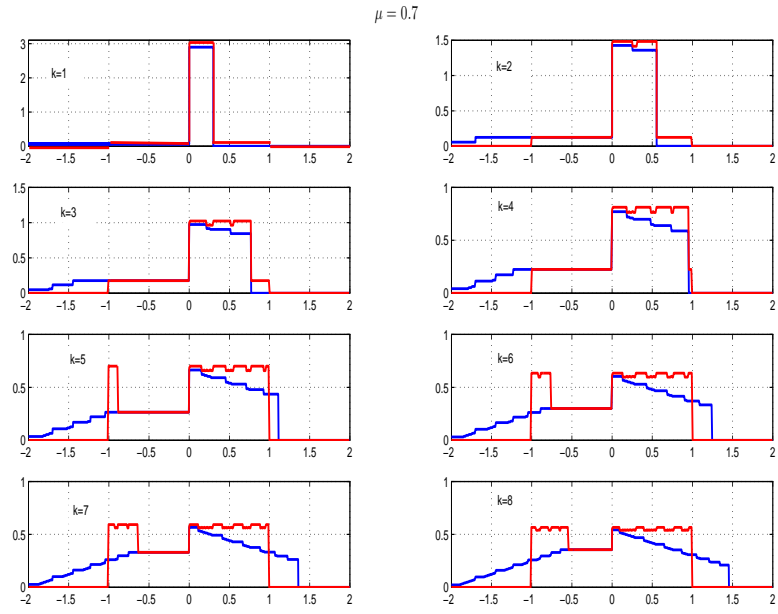


Figure 11: Probability density of $y = T_\mu^k(x) - x$ (Blue line) and \tilde{y} (red line) for $\mu = 0.7$

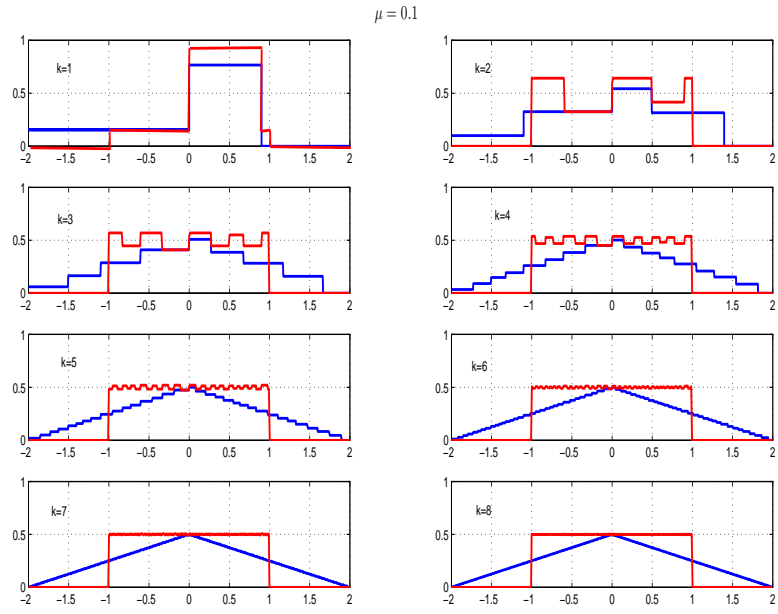


Figure 12: Probability density of $y = T_\mu^k(x) - x$ (Blue line) and \tilde{y} (red line) for $\mu = 0.1$

previous discussion \tilde{y} should be uniform in the interval $[-1, 1]$ in order to have a Rayleigh distribution for $R = |Z_n|$ and consequently to minimize the side lobes of the ambiguity function.

255 We can see that after a few iterations the probability density of $y = T_\mu^k(x) - x$ tends to the triangular distribution in the interval $[-2, 2]$ and thus \tilde{y} follows the uniform distribution in the interval $[-1, 1]$. We did the same thing for $\mu = 0.1$; the results are shown in the Figure 12 . We have almost the same behaviour, with the difference that for the case $\mu = 0.1$ the distribution converges more quickly
260 with respect to the case $\mu = 0.7$. In the general case we obtain a distribution that converges to the triangular one for all values of μ , subsequently we have to look for parameter μ that yields the uniform distribution for the smallest value of iteration k .

To measure the resemblance of $f_{\mu,k}(y)$ with the uniform distribution on the
265 interval $[-1, 1]$ we considered the criteria

$$C(\mu, k) = \int_{-1}^1 (f_{\mu,k}(x) - \frac{1}{2})^2 dx \quad (43)$$

We plotted on Figure 13 $C(\mu, k)$ versus the iteration number k and for different

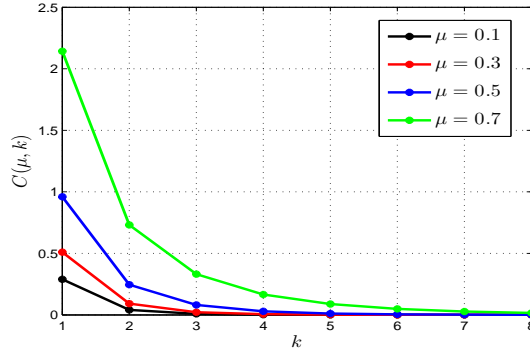


Figure 13: $C(\mu, k)$ for different values of μ and k

values of μ . It is clear that the case $\mu = 0.1$ allows the best value of $C(\mu, k)$.

To see the impact of this result, we plotted on Figure 14 the main term (9) involved in the ambiguity function (5) as a function of ν and for $k = 1, 2, 3$ and

for $\mu = 0.1$ and $\mu = 0.7$. As expected via the above analysis, the case $\mu = 0.1$ allows side lobes in the ambiguity function (9) smaller than the case $\mu = 0.7$.

We have considered here only the first values of k since for higher values

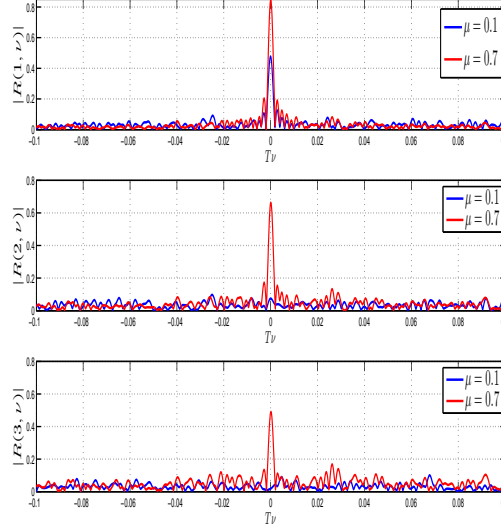


Figure 14: $|R(k, \nu)|$ for $k = 1, 2, 3$; $\mu = 0.1$ and $\mu = 0.7$

$|R(k, \nu)|$ behaves similarly regardless of the bifurcation parameter of the chaotic sequence. Indeed the invariant distribution reaches the uniform distribution for large values of k as explained above. In addition as previously explained for $\nu \neq 0$ the phase coded sequence is distributed on the unit circle giving good behaviour of $|R(k, \nu)|$ even with bad distribution; this explains the similarity of the peaks of $|R(k, \nu)|$ for $\mu = 0.1$ and $\mu = 0.7$

This can also be seen in Figures 15 and 16 where the contours of the ambiguity functions are plotted for $\mu = 0.1$ and $\mu = 0.7$ respectively. We zoomed in on the small values of k and ν because in the other regions, the behaviour of the ambiguity function is almost the same for $\mu = 0.1$ and $\mu = 0.7$. In these Figures and in Figures 20 and 21 below we have removed the maximum peaks of $|R(0, \nu)|$; indeed $|R(0, \nu)|$ is independent of the phase coded sequences (as

285 shown under equations (9)-(10)).

In Figure 17 we have plotted the maximum value of the first side lobe of the

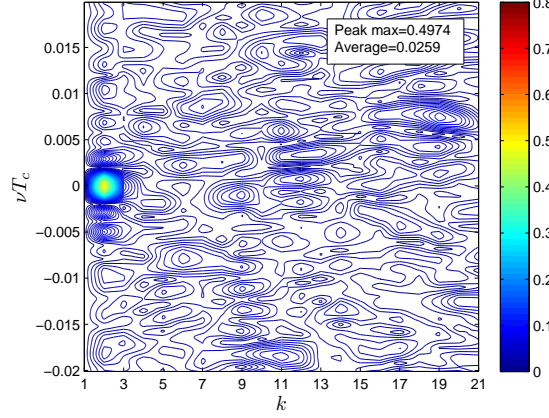


Figure 15: Ambiguity function ($|R(0, \nu)|$ removed) obtained with chaotic waveform ($\mu = 0.1$)

ambiguity function and the Lyapunov exponent (the well-known parameter that characterizes the instability of a dynamic system) with respect to the bifurcation parameter μ . We can see that the more chaotic the sequences, i.e. the larger
 290 the Lyapunov exponent, the lower the maximum value of the first side lobe of the ambiguity function.

It should be noted that other chaotic transformations than the skew tent map could have been envisaged in order to obtain a better distribution of the sequence generated in particular in the first iterations and could thus avoid the
 295 high peaks around zero (that is to say for $k = 1$) in the ambiguity function. If we have considered here the skew tent map as an example for generating chaotic sequences, it is because in our opinion, its theoretical analysis is easier than that of other chaotic transformations. In the next section, we propose a way to construct sequences extracted from the sequences generated by the skew
 300 tent map, producing a reduced first side lobe in the ambiguity function.

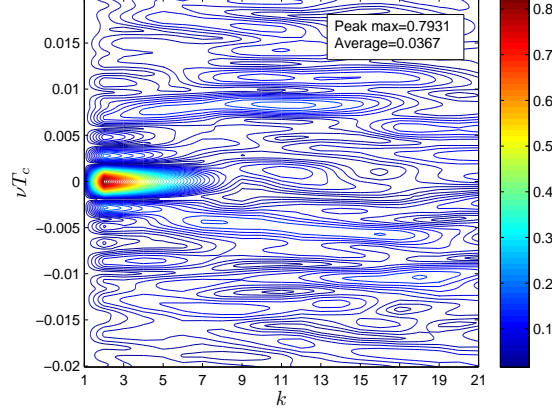


Figure 16: Ambiguity function ($|R(0, \nu)|$ removed) obtained with chaotic waveform ($\mu = 0.7$)

5. Ambiguity function for down-sampled chaotic sequences

We have shown in the previous section how a chaotic sequence generated by the piece-wise linear skew tent map can yield a pseudo-random sequence with a desired probability distribution and, consequently, a good ambiguity function.

305 From the above analysis, however, we have seen that the ambiguity function is not as good as expected for the first values of k because the distribution of the chaotic sequence has not yet converged to the permanent and desired distribution, i.e. the uniform distribution in $[-1, 1]$.

To circumvent this problem we here consider a down-sampled version of the
 310 chaotic sequence, i.e. instead of sequences $\{x_p, 0 \leq p \leq N_c - 1\}$ defined as in section 4 we consider sequences $\{x_p^s, 0 \leq p \leq N_c - 1\}$ defined by: $x_p^s = x_{pN_s}, 0 \leq p \leq N_c - 1$ where N_s is the sampling rate. Noting that $x_{p+k}^s - x_p^s = x_{N_s p + N_s k} - x_{N_s p}$ the probability density of $x_{p+k}^s - x_p^s$ is the same as the one of $x_{q+N_s k} - x_q$. For example, the distributions of $x_{p+k}^s - x_p^s$ for $k = 1, 2$, $N_s = 5$ and when x_p is
 315 generated by (33) with $\mu = 0.1$ are plotted in Figure 18.

Comparing these distributions to the ones in Figures 11 and 12 and for the same values of $k = 1, 2$ we can see that we achieved the desired uniform probability density for all values of k . To analyse the impact of this down-sampling on

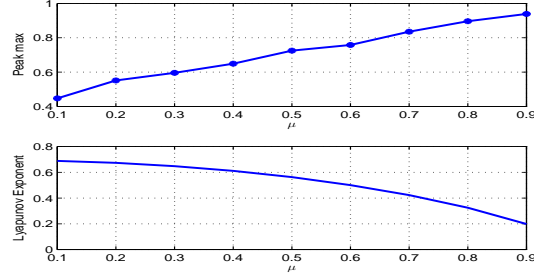


Figure 17: Maximal peak of the ambiguity function ($|R(0, \nu)|$ removed) and Lyapunov exponent versus the bifurcation parameter μ

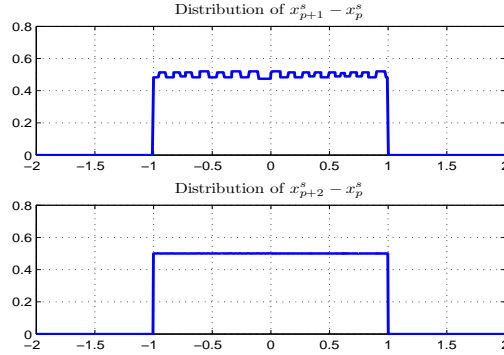


Figure 18: Distribution of $x_{p+1}^s - x_p$ and $x_{p+2}^s - x_p$ for $N_s = 5$

the ambiguity function we plotted on Figure 19 $|R(k, \nu)|$, $k = 1, 2, 3$ for x_p and
320 x_p^S . We can clearly see the superiority of the down-sampled sequence x_p^S with
respect to the sequence x_p .

The resulting ambiguity function in the case $\mu = 0.1$ and for $N_s = 5$ from
which we have removed the maximum peaks of $|R(0, \nu)|$ is shown in Figure 20.
By comparing this ambiguity function to that of Figure 15 corresponding to
325 $\mu = 0.1$ and for $N_s = 1$ we can see the disappearance of the high peaks close
to $(k, \nu) = (1, 0)$ (corresponding to the first side lobe). We also see that the
maximum value of the side lobes is reduced while their average value is almost
the same.

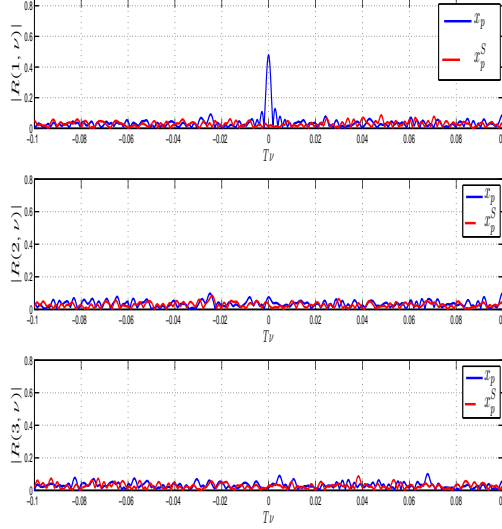


Figure 19: $|R(k, \nu)|, k = 1, 2, 3$ for x_p and x_p^S

We have compared the performances of the waveforms presented here to those proposed in [26]. In this work, the authors generated phase coded waveforms from the so-called cyclic-new algorithm (CAN), an iterative and cyclic algorithm minimizing a cost function based on the ambiguity function. By comparing Figures 20 and 21, and in particular the maximum and average values of the side lobes of the ambiguity functions, it can be seen that the sequences based on the chaos proposed with a down sampling give almost the same performances as CAN sequences. The advantage of chaotic waveforms over CAN waveforms is due to the simplicity of their generation and this regardless of their length or number. Indeed by randomly choosing the initial condition in (33) we obtain a new sequence having the same statistical properties as the others. This also gives the possibility of obtaining sequences with desired spectral characteristics. On the contrary, in the case of CAN waveforms, if we want to add a new waveform and / or a longer waveform to those already generated by CAN, further optimization of the cost function is necessary.

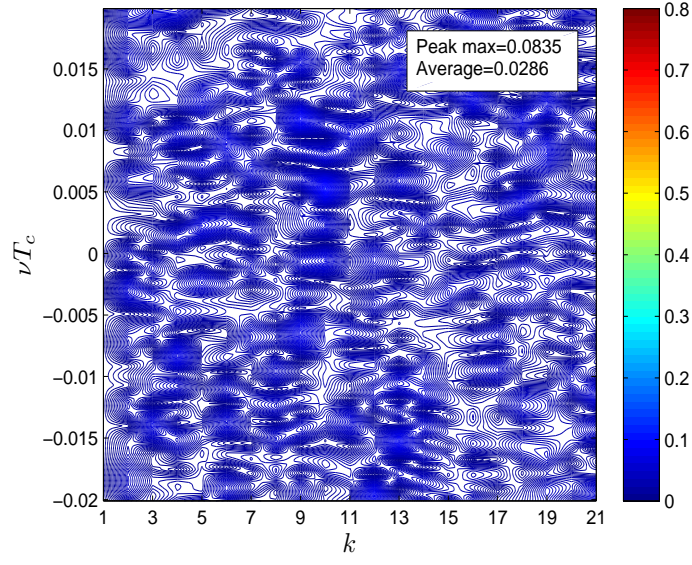


Figure 20: Ambiguity function ($|R(0, \nu)|$ removed) obtained with down-sampled chaotic waveform ($\mu = 0.1$)

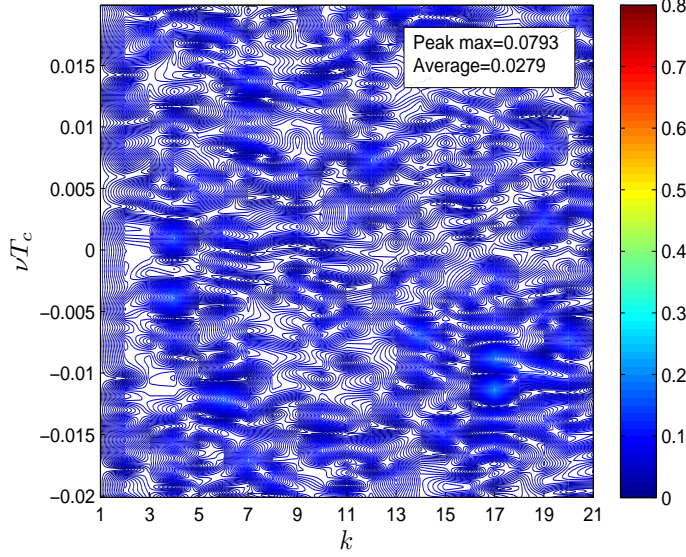


Figure 21: Ambiguity function ($|R(0, \nu)|$ removed) obtained with CAN waveforms

345 6. Conclusion

In this paper we adopted a statistical approach to look for sequences that optimize the ambiguity function of a radar system. We have shown that desired sequences can be generated by a random variable with an appropriate distribution and that it is possible to obtain such sequences by generating them using a chaotic skew tent map. The advantage of using chaotic sequences is that they can be generated of any length and in large numbers. Although generated from a deterministic map with few parameters, they can be considered pseudo-random which makes it possible to study their statistical characteristics. Note that other chaotic sequences could be used and studied.

355 In this paper, we considered the ambiguity function in the case of a basic radar system. It would be easy to demonstrate that the proposed analysis is still valid for the single input-multi-output (SIMO) radar system. The multi-input-multi-output (MIMO) case is however less straightforward since it involves the

generation of multiple sequences with good properties of the cross-correlations
 360 which are embedded in other parameters such that the transmission antenna
 geometry. Because of the promising results we have here obtained, we will ex-
 amine the MIMO case in a future work and adopt the same statistical approach
 to optimize the ambiguity function by considering the cross-correlation functions
 of the sequences.

365 References

- [1] M. A. Richards, Fundamentals of radar signal processing, second edition,
 IET, Mac Graw-Hill Education (2014).
- [2] P. M. Woodward, Probability and information theory, with applications to
 radar, D. W. Fry W. Higinbotham (1953).
- 370 [3] D. Eustice, C. Baylis, R. J. Marks, Woodward’s ambiguity function:
 From foundations to applications, Texas Symposium on Wireless and Mi-
 crowave Circuits and Systems (WMCS) (2015). doi:10.1109/WMCaS.
 2015.7233208.
- [4] M. R. Bell, Information theory and radar waveform design, IEEE Transac-
 375 tions on Information Theory 39 (5) (1993) 1578–1597. doi:10.1109/18.
 259642.
- [5] K. Alhujaili, V. Monga, M. Rangaswamy, Quartic gradient descent for
 tractable radar slow-time ambiguity function (staf) shaping, IEEE Trans-
 actions on Aerospace and Electronic Systems (2019). doi:10.1109/TAES.
 380 2019.2934336.
- [6] H. E. Najafabadi, H. Leung, P. W. Moo, Unimodular waveform design
 with desired ambiguity function for cognitive radar, IEEE Transactions
 on Aerospace and Electronic Systems (2019). doi:10.1109/TAES.2019.
 2942411.

- 385 [7] J. Li, P. Stoica, MIMO radar signal processing, John Wiley and Sons Inc. New Jersey (2009).
- [8] H. Sun, F. Brigui, M. Lesturgie, Analysis and comparison of MIMO radar waveforms, International Radar Conference, Lille, France (13-17 October 2014). doi:10.1109/RADAR.2014.7060251.
- 390 [9] M. Bolhasani, E. Mehrshahi, S. Ghorashi, M. Alijani, Constant envelope waveform design to increase range resolution and SINR in correlated MIMO radar, Signal Processing 163 (2019) 59–65. doi:https://doi.org/10.1016/j.sigpro.2019.05.009.
- [10] P. Stoica, H. He, J. Li, On designing sequences with impulse-like periodic correlation, IEEE Signal Processing Letters 16 (8) (2009) 703–706. doi:395 10.1109/LSP.2009.2021378.
- [11] H. He, P. Stoica, J. Li, Designing unimodular sequence sets with good correlations—including an application to MIMO radar, IEEE Trans. on signal processing 57 (11) (2009) 4391–4405. doi:10.1109/TSP.2009.2025108.
- 400 [12] F. Arlery, R. Kassab, U. Tan, F. Lehmann, Efficient gradient method for locally optimizing the periodic/apperiodic ambiguity function, IEEE Radar Conference (RadarConf), Philadelphia, PA, USA (2-6 May 2016). doi:10.1109/RADAR.2016.7485309.
- [13] U. Tan, al, Phase code optimization for coherent MIMO radar via a gradient descent, IEEE Radar Conference (RadarConf), Philadelphia, PA, USA (2-6405 May 2016). doi:10.1109/RADAR.2016.7485178.
- [14] R. Vignesh, G. A. S. Sundarama, K. P. Somana, Design of less-detectable radar waveforms using stepped frequency modulation and coding, Procedia Computer Science 143 (2018). doi:https://doi.org/10.1016/j.procs.410 2018.10.349.

- [15] A. Aubry, A. D. Maio, B. Jiang, S. Zhang, Ambiguity function shaping for cognitive radar via complex quartic optimization, *IEEE Trans. on signal processing* 61 (22) (2013) 5603–5619. doi:10.1109/TSP.2013.2273885.
- [16] G. Cui, Y. Fu, X. Yu, J. Li, Local ambiguity function shaping via unimodular sequence design, *IEEE Signal Processing Letters* 24 (7) (2017) 977–981. doi:10.1109/LSP.2017.2700396.
- [17] J. Yang, G. Cui, X. Yu, Y. Xiao, L. Kong, Cognitive local ambiguity function shaping with spectral coexistence, *IEEE Access* 6 (2018) 50077–50086. doi:10.1109/ACCESS.2018.2868884.
- [18] M. Khanfouci, S. Marcos, Plm sequences for the performance optimization of linear multiuser detectors, *Eusipco*, Antalya, Turkey (4-8 September 2005).
- [19] S. Merhrzi, S. Marcos, S. Belghith, A family of spatiotemporal chaotic sequences outperforming gold ones in asynchronous ds-cdma systems, *Eusipco*, Florence, Italy (4-8 September 2006).
- [20] Z. B. Jemaa, S. Belghith, Chaotic sequences with good correlation properties for mimo radar application, *SoftCom Conference*, Split, Croatia, (4-8 September 2016). doi:10.1109/SOFTCOM.2016.7772127.
- [21] M. S. Willsey, K. M. Cuomo, A. V. Oppenheim, Quasi-orthogonal wide-band radar waveforms based on chaotic systems, *IEEE Transactions on Aerospace and Electronic Systems* 47 (3) (2011) 1974–1984. doi:10.1109/TAES.2011.5937277.
- [22] Y. Jin, H. Wang, W. Jiang, Z. Zhuang, Complementary-based chaotic phase-coded waveforms design for mimo radar, *IET Radar, Sonar and Navigation* 7 (4) (2013) 371–382. doi:10.1049/iet-rsn.2012.0123.
- [23] P. Beckmann, Statistical distribution of the amplitude and the phase of a multiply scattered field, *Journal of Research of the National Buureau of Standards-Radio Propagation* 66D (3) (1962) 231–240.

- [24] M. Eisencraft, D.M.Kato, L.H.A.Monteiro, Spectral properties of chaotic
 440 signals generated by the skew tent map, Signal Processing 90 (1) (2010)
 385–390. doi:<https://doi.org/10.1016/j.sigpro.2009.06.018>.
- [25] Z. B. Jemaa, D. Fournier-Prunaret, S. Belghith, Kendall’s tau based cor-
 relation analysis of chaotic sequences generated by piecewise linear maps,
 International Journal of Bifurcation and Chaos 25 (13) (2015) 385–390.
 445 doi:DOI:10.1142/S0218127415501771.
- [26] H. He, Waveform design for active sensing systems – a computational
 approach, PhD thesis, University of Florida (2015).

Appendix

Proof of (24) and (25)

450 In this appendix we show (24) where we assumed $a = 2l$, l is a positive
 integer; the demonstration of (25) is almost the same. For simplicity we begin
 by the notations: $W = y_p \in [-a, a]$ and $Z = \tilde{y}_p \in [-1, 1]$. It is obvious that if
 $W = Z + 2k$ where k is an integer, we have $e^{j\pi W} = e^{j\pi Z}$.

We have

$$455 \quad P[Z < z] = 0; \forall z \leq -1 \quad (44)$$

$$P[Z < z] = 1; \forall z \geq 1 \quad (45)$$

For $z \in [-1, 2l - a]$

$$P[Z < z] = P[-1 \leq Z < z] = \sum_{k=1-l}^l P[-1 + 2k \leq W < z + 2k]$$

Knowing that W is uniformly distributed in $[-a, a]$ we have $P[-1 + 2k < W <$
 $z + 2k] = \frac{z+1}{2a}$; we obtain

$$P[Z < z] = 2l \frac{z+1}{2a} = \frac{l}{a}(z+1) \quad (46)$$

For $z \in [2l - a, 0]$

$$P[Z < z] = P[-1 \leq Z < 2l-a] + P[2l-a < Z < z] = \frac{l}{a}(2l-a+1) + \sum_{k=-l}^l P[2l-a+2k < W < z+2k]$$

460 We obtain

$$P[Z < z] = \frac{l}{a}(2l - a + 1) + \frac{2l + 1}{2a}(z + a - 2l) \quad (47)$$

For $z \in [0, a - 2l]$

$$P[Z < z] = P[-1 \leq Z < 0] + P[0 < Z < z]$$

From (47) the first term of this summation is $\frac{1}{2}$, thus

$$P[Z < z] = \frac{1}{2} + \sum_{k=-l}^l P[2k < W < z + 2k]$$

We obtain

$$P[Z < z] = \frac{1}{2} + \frac{2l + 1}{2a}z \quad (48)$$

For $z \in [a - 2l, 1]$

$$P[Z < z] = P[-1 < Z < a - 2l] + P[a - 2l < Z < z]$$

465 From (48) the first term of this summation is $\frac{(2l+1)(a-2l)}{2a}$, thus

$$P[Z < z] = \frac{(2l + 1)(a - 2l)}{2a} + \sum_{k=-l}^{l-1} P[a + 2k - 2l < W < z + 2k]$$

We obtain

$$P[Z < z] = \frac{(2l + 1)(a - 2l)}{2a} + \frac{l}{a}(z - a + 2l) \quad (49)$$

The distribution of Z is obtained by deriving the expressions (45-49) and thus we obtain (24).

Proof of (27) and (28)

470 As the demonstrations of the expressions (27) and (28) are similar we show here only (27).

Let $W = y_p \in [-a, a]$ and $Z = \tilde{y}_p \in [-1, 1]$. It is obvious that if $W = Z + 2k$ where k is an integer, we have $e^{j\pi W} = e^{j\pi Z}$.

Also we have (44) and (45).

475 For $z \in [-1, 2l - a]$

$$P[Z < z] = P[-1 \leq Z < z] = \sum_{k=1-l}^l P[-1 + 2k \leq W < z + 2k]$$

$$f_W(w) = \begin{cases} \frac{w+a}{a^2}, \forall w \in [-a, 0] \\ \frac{-w+a}{a^2}, \forall w \in [0, a] \end{cases}$$

Thus we have

$$P[-1+2k < W < z+2k] = \begin{cases} \int_{-1+2k}^{z+2k} \frac{-w+a}{a^2} dw, & \text{if } k \geq 1 \\ \int_{-1+2k}^{z+2k} \frac{w+a}{a^2} dw, & \text{if } k \leq 0 \end{cases}$$

which gives

$$P[Z < z] = \frac{1}{a^2} \sum_{k=1-l}^0 \left[\frac{1}{2}z^2 + (a+2k)z + C_1 \right] + \frac{1}{a^2} \sum_{k=1}^l \left[-\frac{1}{2}z^2 + (a-2k)z + C_2 \right]$$

where C_1 and C_2 are constants; By simplifying we obtain

$$P[Z < z] = \frac{1}{a^2}(2al - l^2)z + C_1 + C_2 \quad (50)$$

480 For $z \in [2l - a, 0]$

$$P[Z < z] = P[Z < 2l - a] + P[2l - a \leq Z < z]$$

$$P[Z < z] = C + P[2l - a \leq Z < z]$$

C is constant

$$P[Z < z] = C + \sum_{k=-l}^l P[2l - a + 2k \leq W < z + 2k]$$

$$P[Z < z] = C + \frac{1}{a^2} \sum_{k=-l}^0 \left[\frac{1}{2}z^2 + (a+2k)z \right] + C_3 + \frac{1}{a^2} \sum_{k=1}^l \left[-\frac{1}{2}z^2 + (a-2k)z \right] + C_4$$

which gives

$$P[Z < z] = \frac{1}{a^2} \left[\frac{1}{2}z^2 + (2l+1)az - 2l(l+1)z \right] + C + D$$

485 $D = C_3 + C_4$ Thus,

$$P[Z < z] = \frac{1}{a^2} \left[\frac{1}{2}z^2 + ((2l+1)a - 2l(l+1))z \right] + C + D \quad (51)$$

By deriving (50) and (51) we find the distribution of Z for $z < 0$, for the symmetry of the problem we deduce the distribution of Z for $z > 0$; thus we find

$$f_Z(z) = \begin{cases} \frac{2al-l^2}{a^2}, & \text{if } a-2l \leq |z| \leq 1 \\ \frac{(2l+1)a-2l(l+1)-|z|}{a^2}, & \text{if } 0 < |z| \leq a-2l \end{cases}$$